# Control of active suspension system using $H_{\infty}$ and adaptive robust controls

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## ABSTRACT:

This paper presents a control of active suspension system for quarter-car model with two-degree-of-freedom using  $H_{\infty}$  and nonlinear adaptive robust control method. Suspension dynamics is linear and treated by  $H_{\infty}$  method which guarantees the robustness of closed loop system under the presence of

uncertainties and minimizes the effect of road disturbance to system. An Adaptive Robust Control (ARC) technique is used to design a force controller such that it is robust against actuator uncertainties. Simulation results are given for both frequency and time domains to verify the effectiveness of the designed controllers.

**Keywords:** Active suspension, Hydraulic actuator,  $H_{\infty}$  control, Adaptive robust control.

#### **1. INTRODUCTION**

Automotive suspension systems have been developed from the begin time of car industrial with a simple passive mechanism to the present with a very high level of sophistication. Suspensions incorporating active components are studied to improve the overall ride performances of automotive vehicle in recent years. Active suspension must provide a trade-off between several competing objectives: passenger comfort, small suspension stroke for packing and small tire deflection for vehicle handling. In the early studies, linear model of suspension are used with the assumption of ideal force actuator. The most applicable force actuator using in practice is hydraulic actuator that has a high non-linearity characteristic. Hence to solve completely problem, recently studies consider to the dynamics and the non-linearity of hydraulic actuator <sup>[2,7,9]</sup>.

This paper presents a control of active suspension system for quarter-car model with two-degree-of-freedom by using  $H_{\infty}$  and nonlinear adaptive robust control method. The

system is divided into two parts: the linear part is whole system except actuator and nonlinear part is hydraulic actuator. The linear part is treated using  $H_{\infty}$  control method that guarantees the robustness of closed loop system under the presence of uncertainties and minimizes the effect of disturbance. The variations of system parameters are solved by multiplicative uncertainty model. In hydraulic actuator, there are some unknown factors such as bulk modulus of hydraulic fluid that has strong effect to actuator dynamics. Hence, the nonlinear adaptive control is suitable for designing actuator controller. This paper applied the ARC technique to design a the controller robust against actuator uncertainties<sup>[3,4]</sup>. The error between desired acting force calculated from  $H_{\infty}$  controller and actual force generated by hydraulic actuator is considered as the disturbance to the linear system. Simulations have been done in both frequency and time domains to verify the effectiveness of the designed controllers.

#### 2. SYSTEM MODELING

The scheme of suspension system and hydraulic actuator used in this paper is described in Fig. 1.





Fig.1 Suspension system and actuator

Define parameters as the follows

- $m_s$  : sprung mass
- $m_{\mu}$  : unsprung mass
- $b_s$  : damping coefficient
- $k_{s}$  : spring stiffness coefficient
- $k_t$  : tire stiffness coefficient
- F : active force
- $z_s$  : displacement of the car body
- $z_u$  : displacement of wheel
- $z_r$  : displacement of road

Assume that the spring stiffness coefficient and tire stiffness coefficient are linear in their operation range; the tire does not leave the ground; and  $z_s$  and  $z_u$  are measured from the static equilibrium point. From the scheme of the system model in the Fig. 1, the state variables are chosen as follows

$$x_1 = z_s - z_u$$
: suspension deflection  
 $x_2 = x_s$ : velocity of car

body

$$x_3 = z_u - z_r$$
: tire deflection

$$x_4 = \mathcal{K}_{a}$$
 : velocity of wheel  
 $x_5 = F$  : active force

 $x_6 = x_{valve}$  : position of valve from its closed position.

The governing dynamic equations of suspension system including hydraulic actuator can be presented as the following  $^{[9]}$ 

$$\mathbf{k} = x_2 - x_4 \tag{1}$$

$$\mathbf{x}_{2} = \frac{1}{m_{s}} \left( -k_{s} x_{1} - b_{s} (x_{2} - x_{4}) + x_{5} \right)$$
(2)

$$\mathbf{x}_{3} = x_{4} - \mathbf{x}_{7} \tag{3}$$

$$\mathbf{x}_{4} = \frac{1}{m_{u}} \left( k_{s} x_{1} + b_{s} \left( x_{2} - x_{4} \right) - k_{t} x_{3} - x_{5} \right)$$
(4)
$$\mathbf{x}_{4} = -\mathbf{B} x_{1} - \mathbf{A} x_{2} \left( x_{1} - x_{1} \right) + \mathbf{A} x_{3} - \mathbf{A} x_{5} + \mathbf{A$$

$$\mathbf{x}_{5} = -\beta x_{5} - \alpha_{f} A (x_{2} - x_{4}) + \gamma \sqrt{A} \sqrt{P_{s} A - \operatorname{sgn}(x_{6}) x_{5}} x_{6}$$
(5)

$$\mathbf{x}_{6} = \frac{1}{\tau} (-x_{6} + u) \tag{6}$$

where,

$$\begin{split} \gamma &= \alpha_f \ C_d \ w_f \ \sqrt{1/\rho} \\ \beta &= \alpha_f \ C_{im} \\ \alpha_f &= 4 \beta_e \ / V_t \\ A &: \text{piston area} \\ P_s &: \text{supply pressure of the fluid} \\ C_d &: \text{discharge coefficient} \\ w_f &: \text{spool valve area gradient} \\ \rho &: \text{hydraulic fluid density} \\ C_{im} &: \text{total leakage coefficient of the piston} \\ \beta_e &: \text{effective bulk modulus} \\ V_t &: \text{total actuator volume} \\ \tau &: \text{time constant} \\ u &: \text{input to servo-valve} \\ \text{Equations (1)-(4) represent the quarter-care} \end{split}$$

dynamics and equations (5)-(6) drive the

hydraulic actuator dynamics.

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# 3. $H_\infty$ control of linear part

Let's define the force error

$$e = x_5 - x_5^{\ d} \tag{7}$$

where  $x_5$  is actual control force generated from actuator and  $x_5^d$  is the desired control force which is calculated from  $H_{\infty}$  controller. Consider  $x_5$  as the control input, the systems (1)-(4) can be rewritten in the form

$$\mathbf{x}_{p} = A_{p} x_{p} + B_{p} x_{5} + \Gamma \begin{bmatrix} \mathbf{x}_{p} \\ e \end{bmatrix}$$

and the measured output is the velocity of car body

$$y_p = C_p x_p \tag{9}$$

where

$$\begin{aligned} x_{p} &= \begin{bmatrix} x_{1} \\ x_{2} \\ x_{3} \\ x_{4} \end{bmatrix}, \\ A_{p} &= \begin{bmatrix} 0 & 1 & 0 & -1 \\ -\frac{k_{s}}{m_{s}} & -\frac{b_{s}}{m_{s}} & 0 & \frac{b_{s}}{m_{s}} \\ 0 & 0 & 0 & 1 \\ \frac{k_{s}}{m_{u}} & \frac{b_{s}}{m_{u}} & -\frac{k_{t}}{m_{u}} & -\frac{b_{s}}{m_{u}} \end{bmatrix}, \\ B_{p} &= \begin{bmatrix} 0 \\ \frac{1}{m_{s}} \\ 0 \\ -\frac{1}{m_{u}} \end{bmatrix}, \qquad \Gamma = \begin{bmatrix} 0 & 0 \\ 0 & \frac{1}{m_{s}} \\ -1 & 0 \\ 0 & -\frac{1}{m_{u}} \end{bmatrix}, \\ C_{p}^{T} &= \begin{bmatrix} 0 \\ 1 \\ 0 \\ 0 \end{bmatrix}$$

Three interest performance variables are: body vibration isolation, measured by the sprung mass acceleration  $\mathbf{E}_{\mathbf{x}}$ ; suspension travel, measured by the deflection of suspension  $z_s - z_u$ ; and tire load constancy, measured by the tire deflection  $z_u - z_r$ . Then three considered transfer functions from disturbance  $\mathbf{E}_{\mathbf{x}}$  to the acceleration of the sprung mass  $H_A(s)$ , to the suspension deflection  $H_{SD}(s)$ , and to the tire deflection  $H_{TD}(s)$  can be derived as the following

$$H_{A}(s) = \frac{\cancel{B}_{s}(s)}{\cancel{B}_{r}(s)} = \frac{\cancel{B}_{2}(s)}{\cancel{B}_{r}(s)}$$
(10)

$$Z_s(s)$$

$$H_{SD}(s) = \frac{Z_{s}(s) - Z_{u}(s)}{Z_{r}(s)} = \frac{X_{1}(s)}{Z_{r}(s)}$$

(11)

(8)

$$H_{TD}(s) = \frac{Z_u(s) - Z_r(s)}{Z_r(s)} = \frac{X_3(s)}{Z_r(s)}$$
(12)

The augmented system G(s) for  $H_{\infty}$  control problem is given in the Fig. 2.



#### Fig. 2. Configuration of control system

The state space expression of the plant P(s) with adding measurement noise n can be written in the following form

$$\mathbf{x}_{p} = A_{p} x_{p} + B_{p1} w + B_{p2} x_{5}$$

$$z_{p} = C_{p1} x_{p} + D_{p11} w + D_{p12} x_{5}$$
(13)
(14)

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$$y_p = C_{p2}x_p + D_{p21}w + D_{p22}x_5$$
(15)

The state space expression of the plant G(s) can be written as follows

$$\mathbf{k} = Ax + B_1 w + B_2 x_5 \tag{16}$$

$$z = C_1 x + D_{11} w + D_{12} x_5 \tag{17}$$

$$y = C_2 x + D_{21} w + D_{22} x_5 \tag{18}$$

where,

$$x = \begin{bmatrix} x_p \\ x_w \end{bmatrix}, z = z_p, y = y_p,$$

$$A = \begin{bmatrix} A_p & 0 \\ B_w C_{p11} & A_w \end{bmatrix}$$

$$B_1 = \begin{bmatrix} B_{p1} \\ B_w D_{p111} \end{bmatrix}, B_2 = \begin{bmatrix} B_{p2} \\ B_w D_{p121} \end{bmatrix},$$

$$C_1 = \begin{bmatrix} D_w C_{p11} & B_w \\ \alpha_w D_{p12} & 0 \end{bmatrix},$$

$$C_2 = \begin{bmatrix} C_{p2} & 0 \end{bmatrix}$$

$$D_{11} = \begin{bmatrix} D_w D_{p111} \\ \alpha_w D_{p112} \end{bmatrix}, D_{12} = \begin{bmatrix} D_w D_{p121} \\ \alpha_w D_{p122} \end{bmatrix}$$

$$D_{21} = D_{p21}, D_{22} = D_{p22}$$

The  $H_{\infty}$  control problem is to find an internal stabilizing controller, K(s), for the augmented system, G(s), such that the inf-norm of the closed loop transfer function,  $T_{zw}$ , is below a given positive scalar  $\gamma$ 

Find 
$$\|T_{zw}\|_{\infty} \le \gamma$$
  
 $K(s)$ stabilizing (19)

Furthermore, from the small gain theorem the robust stability of the closed loop system under presence of parameter uncertainty is assured if  $\gamma < 1$ . Here the change of the parameters of the system is treated by multiplicative uncertainty

model  $\Delta(s)$ . It is derived from the nominal plant  $P_n(s)$  and the perturbed plant  $P_p(s)$  as follows

$$\Delta(s) = \frac{P_p(s)}{P_n(s)} - 1$$
(20)

The weighting is chosen to satisfy

$$\sigma[\Delta(s)] < |W_1(s)|, \quad \forall \omega \tag{21}$$

The transfer function from disturbance to the state of the augmented system is

$$T_{x_{\frac{2}{2}}} = \left\{ sI - [A + B_2 K(s)C_2] \right\}^{-1} [B_1 + B_2 K(s)D_{21}] \begin{bmatrix} 1\\ 0\\ 0 \end{bmatrix}$$
(22)

where K(s) is  $H_{\infty}$  controller. Three transfer functions (10)-(12) become

$$H_{AC}(s) = s \begin{bmatrix} E_2 & 0 \end{bmatrix} T_{x \text{ s} \text{ s}}$$
$$H_{SD}(s) = \begin{bmatrix} E_1 & 0 \end{bmatrix} T_{x \text{ s} \text{ s}}$$
$$H_{TD}(s) = \begin{bmatrix} E_3 & 0 \end{bmatrix} T_{x \text{ s} \text{ s}}$$
where

$$E_1 = \begin{bmatrix} 1 & 0 & 0 & 0 \end{bmatrix},$$
  

$$E_2 = \begin{bmatrix} 0 & 1 & 0 & 0 \end{bmatrix}, \quad E_3 = \begin{bmatrix} 0 & 0 & 1 & 0 \end{bmatrix}$$

# 4. ADAPTIVE ROBUST CONTROL OF NONLINEAR PART

In this part we will derive the controller for hydraulic actuator used in suspension system. The controller is designed based on adaptive robust control technique proposed by Bin Yao<sup>[3]</sup>. Consider hydraulic actuator dynamic equations (5)-(6). The parameter is considered as unknown parameter  $\alpha_f = 4\beta_e/V_t$ . The main reason for choosing  $\alpha_f$  as unknown factor is that the bulk modulus of hydraulic fluid is known to change dramatically even when there is a small leakage between piston and cylinder.

The equation (5) can be written in the form

$$\mathbf{x}_{5} = \theta[a_{1}x_{5} + a_{2}(x_{2} - x_{4}) + a_{3}\sqrt{P_{s}A - \text{sgn}(x_{6})x_{5}}x_{6}] + d$$
(23)

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where 
$$a_1 = -C_{tm}$$
;  $a_2 = -A^2$ ;  
 $a_3 = C_d w_f \sqrt{A/\rho}$ ; and  $\theta$  is unknown  
parameter;  $d$  denotes disturbances and  
their extents are known

 $\theta \in \Omega_{\theta} \stackrel{\Delta}{=} \{ \theta : \theta_{\min} < \theta < \theta_{\max} \}$ 

 $|d| < d_M$ 

The adaptive control law can be obtained as the following steps.

Step 1: Let's define

$$b = a_3 \sqrt{P_s A - \operatorname{sgn}(x_6) x_5}$$
(24)

Equation (23) becomes

$$\mathbf{x}_{5} = \theta[a_{1}x_{5} + a_{2}(x_{2} - x_{4}) + bx_{6}] + d$$
(25)

Define the error variable:

$$z_1 \equiv x_5 - x_5^d \tag{26}$$

To find a virtual control law  $\alpha$  for  $x_6$  such that  $x_5$  tracks its desired value  $x_5^d$  using the procedure suggested in [3]. The term b, representing the nonlinear static gain between the flow rate and the valve opening  $x_6$ , is a function of  $x_6$  and also is non-smooth since  $x_6$  appears through a discontinuous function  $\text{sgn}(x_6)$ . So a smooth modification is needed<sup>[3]</sup>.

Define the smooth projection  $\pi(\hat{\theta})$ :

$$\pi(\hat{\theta}) = \begin{cases} \theta_{\max} + \varepsilon_{\theta} \left\{ 1 - \exp\left[ -\frac{1}{\varepsilon_{\theta}} (\hat{\theta} - \theta_{\max}) \right] \right\} & (\hat{\theta} > \theta_{\max}) \\ \hat{\theta} & (\hat{\theta} \in [\theta_{\min}, \theta_{\max}]) \\ \theta_{\min} + \varepsilon_{\theta} \left\{ 1 - \exp\left[ \frac{1}{\varepsilon_{\theta}} (\hat{\theta} - \theta_{\min}) \right] \right\} & (\hat{\theta} < \theta_{\min}) \end{cases}$$

The control law  $\alpha$  is given by

$$\alpha = \alpha_a + \alpha_r \tag{27}$$

The adaptive part  $\alpha_a$  and the robust control part  $\alpha_r$  are calculated as follows

$$\alpha_{a} = \frac{1}{a_{3}} \left[ -a_{1}x_{5} - a_{2}(x_{2} - x_{4}) + \frac{1}{\hat{\theta}_{\pi}} (\mathbf{x}_{5d} - k_{1}z_{1}) \right]$$
(28)
$$\alpha_{r} = -\frac{1}{4\theta_{\min}a_{3}} z_{1} \left[ \frac{1}{\varepsilon_{11}} \theta_{M}^{2} (a_{1}x_{5} + a_{2}(x_{2} - x_{4}) + a_{3}\alpha_{a})^{2} + \frac{1}{\varepsilon_{12}} d_{M}^{2} \right]$$
(29)

where

$$k_1$$
 : tunable parameter

$$\hat{\theta}_{\pi} = \pi(\hat{\theta})$$
$$\theta_{M} = \theta_{\max} - \theta_{\min} + \varepsilon_{\theta}$$

 $\boldsymbol{\theta}$  is estimated by  $\boldsymbol{\theta}$  using the following adaptation law

$$\hat{\theta} = \gamma_1 z_1 [a_1 x_5 + a_2 (x_2 - x_4) + a_3 \alpha_a],$$
  
$$\gamma_1 > 0$$
(30)

 $\mathcal{E}_{\theta}$  is a known arbitrary small positive number

and  $\mathcal{E}_{11}$ ,  $\mathcal{E}_{12}$  are adjustable small positive numbers.

Step 2: To find an actual control law for u

such that  $x_6$  tracks the desired control function  $\alpha$  synthesized in step 1 with a guaranteed transient performance.

Define the error variable

$$z_2 \equiv x_6 - \alpha$$

Adaptive robust control law consists of two parts: an adaptive part and a robust control part

(31)

$$u = u_a + u_r \tag{32}$$

The adaptive part and robust control part are calculated as follows

$$u_{a} = \frac{\tau}{b} \left[ -k_{2}z_{2} - p_{e} - \frac{\partial \alpha}{\partial \hat{\theta}} \gamma_{1}\tau_{2c} \right]$$
(33)

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(34)

$$u_r = -\frac{\tau}{4b} z_2 h_2$$

where

$$p_e = \hat{\theta}_{\pi} \frac{w_1}{w_2} z_1 - \frac{1}{\tau} b x_6 + \frac{\partial b}{\partial x_5} \hat{X}_5 - a_c^{\text{(35)}}$$

$$\hat{X}_{5} = \hat{\theta}_{\pi}[a_{1}x_{5} + a_{2}(x_{2} - x_{4}) + bx_{6}]$$
(36)

$$\mathbf{a}_{c}^{\mathbf{x}} = \frac{\partial \alpha}{\partial x_{5}} \mathbf{a}_{5}^{\mathbf{x}} + \frac{\partial \alpha}{\partial t}$$
(37)

$$\tau_{2c} = \tau_{1c} - w_2 z_2 \Phi \tag{38}$$

$$\tau_{1c} = -w_1 z_1 [a_1 x_5 + a_2 (x_2 - x_4) + a_3 \alpha]$$
(39)

$$\Phi = \frac{w_1}{w_2} z_1 + \left[\frac{\partial b}{\partial x_5} - \frac{\partial \alpha}{\partial x_5}\right] \left[a_1 x_5 + a_2 (x_2 - x_4) + a_3 \alpha\right]$$
(40)

$$h_2 = \frac{1}{\varepsilon_2} \theta_M^2 \Phi^2 \tag{41}$$

 $k_2$ ,  $w_1$ ,  $w_2$  and  $\varepsilon_2$  are arbitrary positive numbers.

# 5. SIMULATION RESULTS

The numerical values using in this simulation are given in the Table  $1^{[9]}$ .

The weighting function is chosen as

$$W(s) = \begin{bmatrix} W_1(s) & 0 \\ 0 & \alpha_W \end{bmatrix} = \begin{bmatrix} \frac{3.135s + 9.2625}{0.93s + 29} & 0 \\ 0 & 3.5 \times 10^{-4} \end{bmatrix}$$

The controller is calculated with the value of  $\gamma = 0.99$ . The road velocity disturbance is assumed to be from road displacement  $r = 0.1 \sin 2\pi f t$ . The parameters of ARC controller are chosen to be  $\gamma_1 = 5e6$ ,  $k_1 = 150$ ,  $k_2 = 10$ ,  $\varepsilon_{\theta} = 0.001$ ,  $\varepsilon_{11} = 5$ ,  $\varepsilon_{12} = 2$ ,  $\varepsilon_2 = 5$  and  $d_M = 2$ .

Fable 1.	Numerical	values	for	simulat	ion

Parameters	Values	Units
m <sub>s</sub>	290	kg
m <sub>u</sub>	59	kg
$b_s$	1000	Ns/m
k <sub>s</sub>	16812	N/m
k,	190000	N/m
$lpha_{_f}$	4.515e13	$N/m^5$
β	1.00	
γ	1.545e9	$N/(m^{5/2}kg^{1/2})$
A	3.35e-4	$m^2$
$P_s$	10342500	N/m <sup>2</sup>

#### **Frequency domain**

The plot of uncertainties and weighting functions are given in Fig. 3. Figures (4)-(6) show the gain plots for three transfer functions (10)-(12) in cases of passive system, active system with desired force and actual force input. As shown in the figures, the designed nonlinear ARC controller can treat the nonlinearity and keep the  $H_{\infty}$  frequency performance well.



Fig. 3. Plots of uncertainties and weighting function



Fig. 4. Gain plots for body acceleration transfer function



Fig. 5. Gain plots for suspension deflection transfer function



Fig. 6. Gain plots for tire deflection transfer function



Fig. 7. Acceleration with step disturbance



Fig. 8. Suspension deflection with step disturbance



Fig. 9. Tire deflection with step disturbance



Fig. 10. Acceleration with sine disturbance



Fig. 11. Suspension deflection with sine disturbance

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**Fig. 12.** Tire deflection with sine disturbance

### Time domain

The responses of the system with step and sine wave disturbances are considered. Responses of the system in case of step disturbance are given in Figs. (7)-(9). The step road velocity is of 0.1 m/s. Body acceleration and tire deflection are much reduced but the suspension deflection is higher. Responses of the

system in case of sine wave disturbance are given in Figs. (10)-(12). The road amplitude is assumed to be 0.1m with frequency of 1Hz. At this frequency, active system reduces considerably the effects of disturbance.

#### 6. CONCLUSION

This paper presents a control of active suspension system using  $H_{\infty}$  and nonlinear adaptive robust control method.  $H_{\infty}$  controller achieved the robustness with the presence of parameter uncertainties and minimized the effects of disturbance. The nonlinear ARC controller treats well the non-linearity and the parameter uncertainties of hydraulic actuator. Simulation results show that the designed controller can keep the good performance of  $H_{\infty}$  controller in both frequency and time domains.

# Điều khiển hệ thống treo chủ động của xe ô tô dùng $H_{\infty}$ và điều khiển thích nghi bền vững

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# TÓM TẮT:

Điều khiển hệ thống treo chủ động của xe ô tô là một đề tài thú vị trong lĩnh vực nghiên cứu về ô tô. Bài báo này đề xuất phương pháp điều khiển hệ thống treo chủ động bằng lý thuyết  $H_{\infty}$  và điều khiển thích nghi bền vững. Kỹ thuật điều khiển thích nghi bền vững (ARC) được sử dụng để thiết kế bộ điều khiển lực bền vững với các thông số không biết chắc của bộ chấp hành. Kết quả mô phỏng đã thể hiện tính hiệu quả của bộ điều khiển đề nghị.

**Từ khóa: :** Hệ thống treo chủ động, Điều khiển  $H_{\infty}$ , Điều khiển thích nghi bền vững.

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